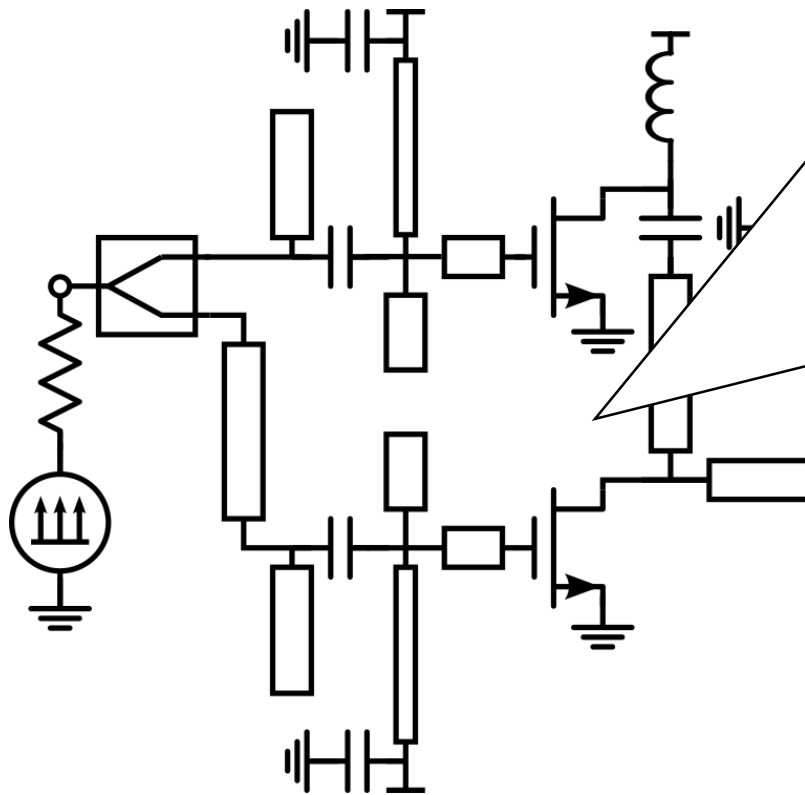


Local stability analysis of microwave circuits

Adam Cooman, Fabien Seyfert, Laurent Baratchart,
Martine Olivi, Sylvain Chevillard



Modern circuit simulation tools



Frequency domain
simulation methods



ADS



MWO

cādence®

Spectre

Proprietary simulator

Proprietary models

Equilibrium Solution

DC simulation

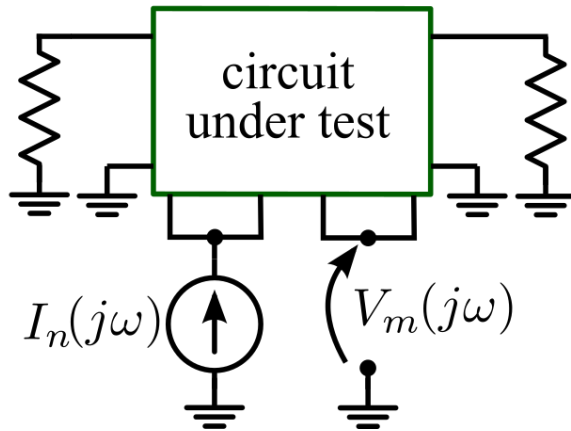
Periodic solution

Harmonic Balance

Local stability analysis: Linearise

Equilibrium Solution

DC simulation

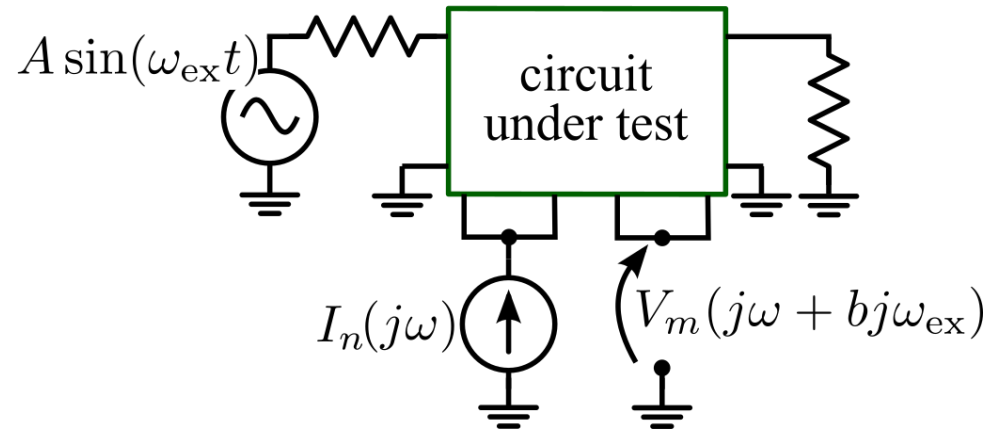


$$Z_{mn}(j\omega) = \frac{V_m(j\omega)}{I_n(j\omega)}$$

AC analysis

Periodic solution

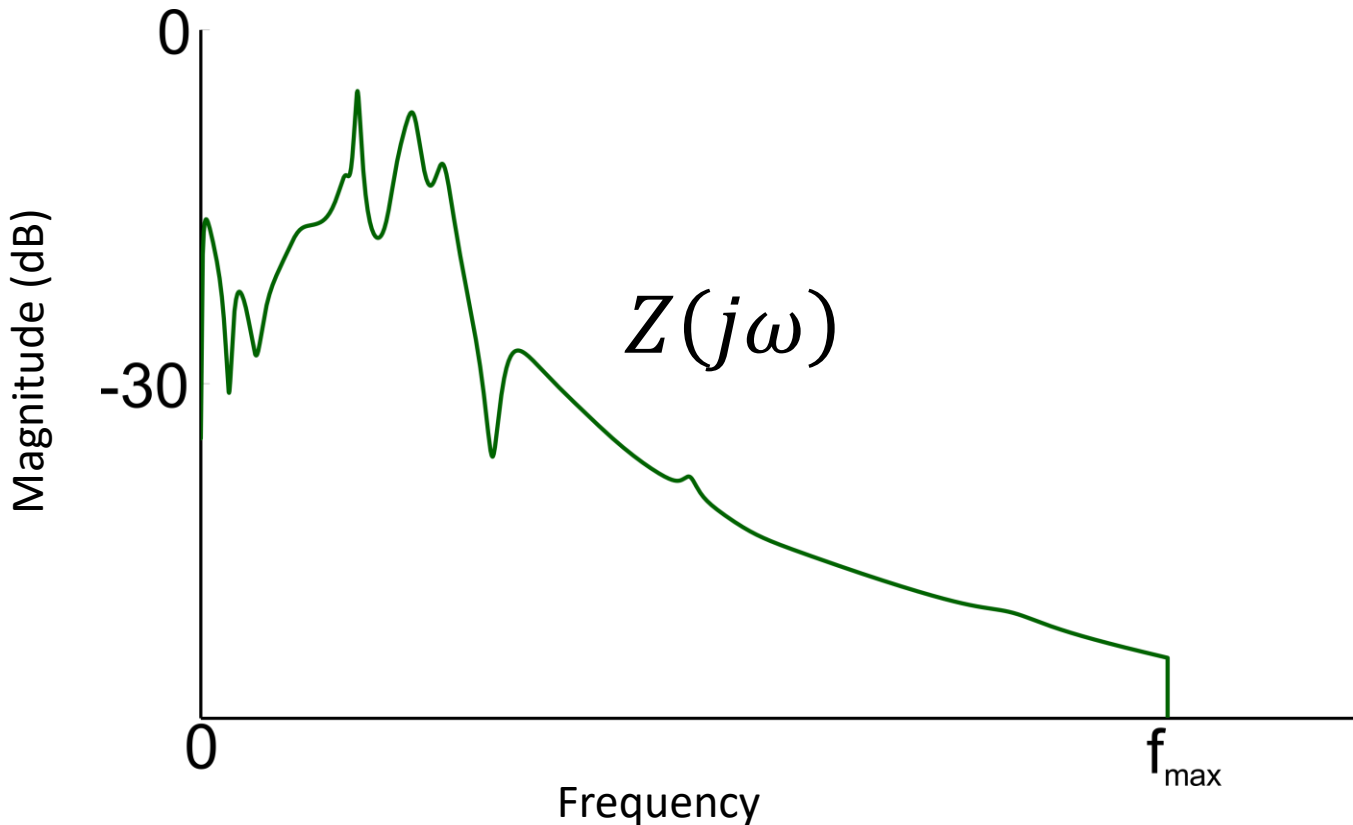
Harmonic Balance



$$Z_{mn}(j\omega) = \frac{V_m(j\omega + bj\omega_{ex})}{I_n(j\omega)}$$

LSSS analysis

Is frequency response stable?



$Z(j\omega)$ known on a discrete set of frequencies
Does $Z(j\omega)$ have right half-plane poles?

Content

Stability Analysis by projection

Examples

Filter influence

Estimating unstable poles

Assumptions:

Circuit contains delay $\Rightarrow Z(j\omega)$ meromorphic

Circuit is realistic \Rightarrow unstable part rational

Unstable pole observable in $Z(j\omega)$

$Z(j\omega) \in \mathcal{L}_2$ \Rightarrow No poles on the $j\omega$ Axis

Noiseless data

Stability in Hardy context

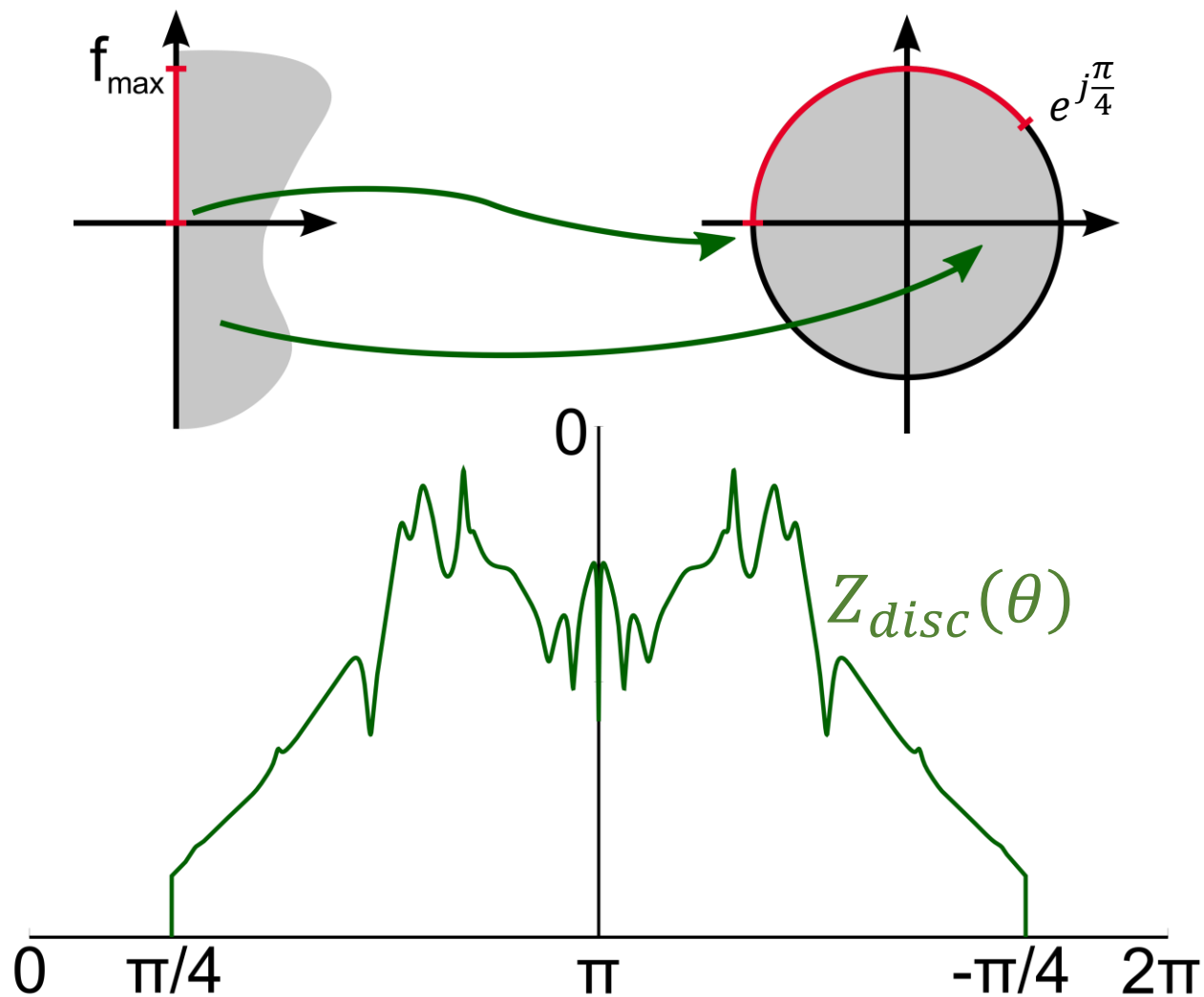
$$g \in \mathcal{H}_2 \text{ when } \begin{cases} g \text{ analytic in RHP} \\ \int |g(j\omega + \sigma)|^2 d\omega < \infty \quad \sigma \rightarrow 0 \end{cases}$$

$$\mathcal{L}_2 = \mathcal{H}_2 \oplus \overline{\mathcal{H}_2}$$
$$Z(j\omega) = Z_{stable}(j\omega) + Z_{unstable}(j\omega)$$

$$Z_{stable}(j\omega) = P_{\mathcal{H}_2}\{Z(j\omega)\}$$

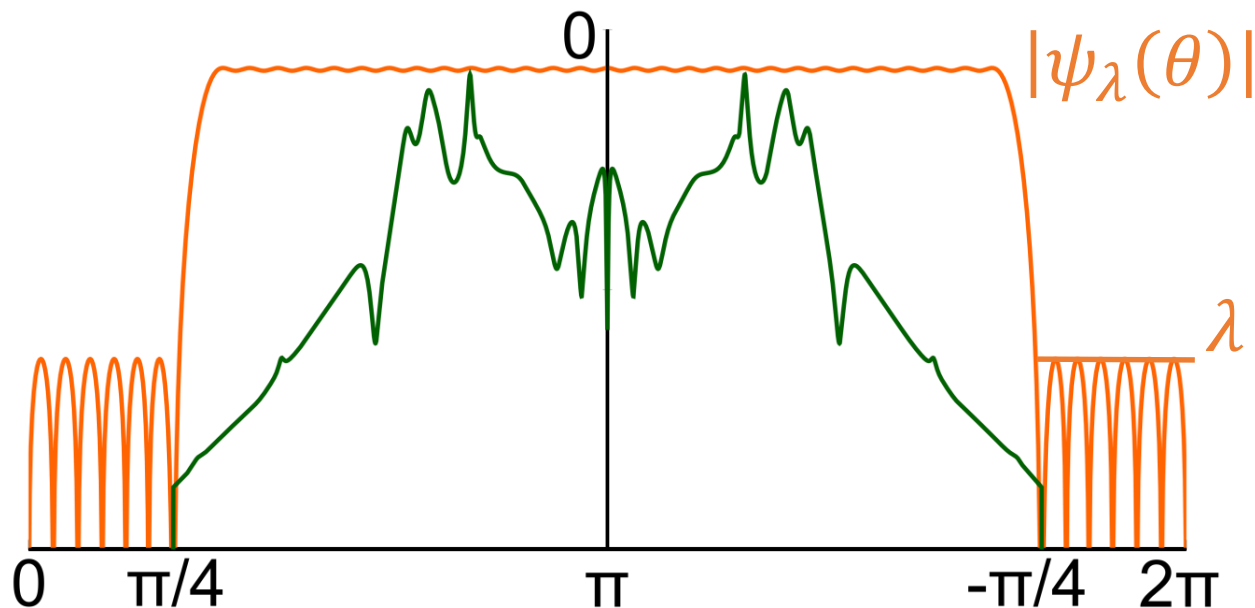
$$Z_{unstable}(j\omega) = P_{\overline{\mathcal{H}_2}}\{Z(j\omega)\}$$

Step 1: Transform to unit circle



Step 2: Multiply by Filter function

$$Z_f = Z_{disc} \cdot \psi_\lambda$$

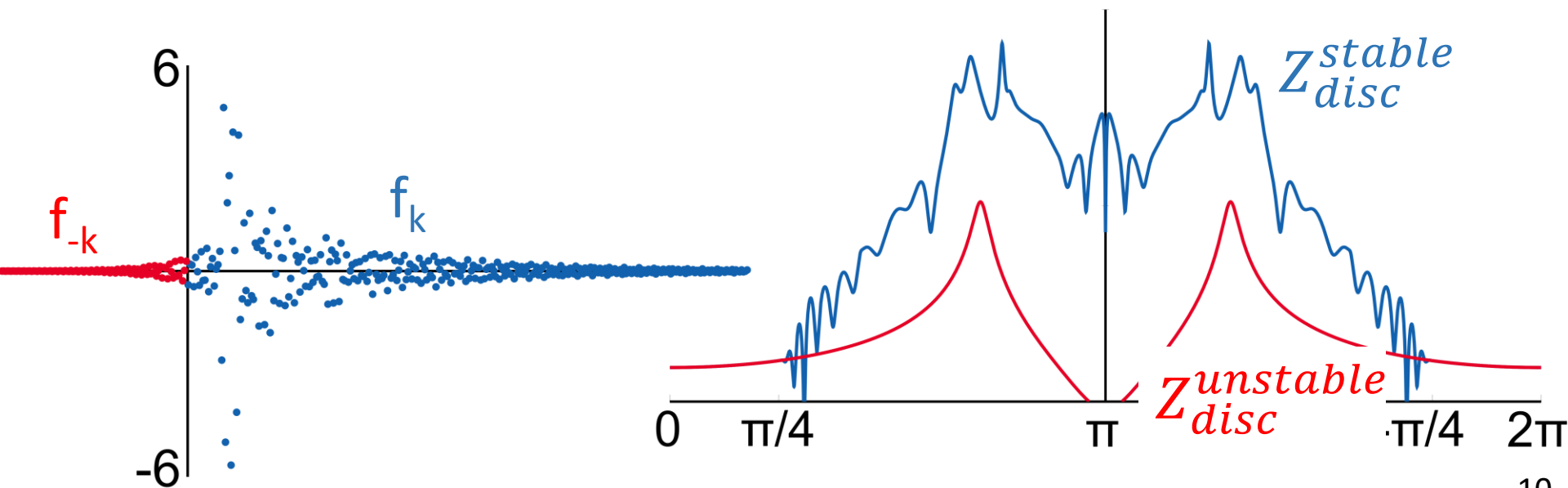


To smooth out edges

To suppress influence of out-of-band data

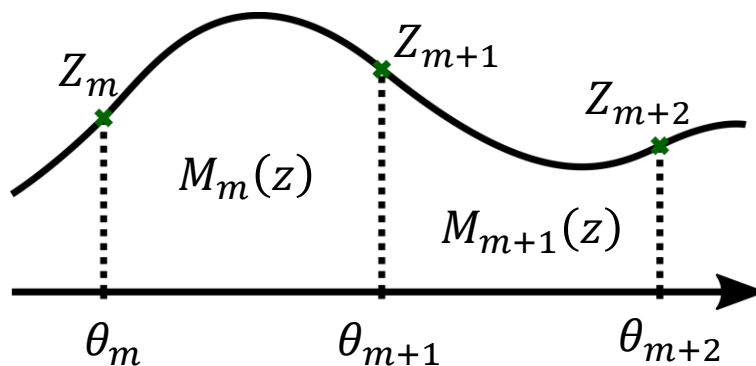
Step 3: Compute Fourier series

$$Z_f(\theta) = \underbrace{\sum_{k=0}^{\infty} f_k e^{jk\theta}}_{Z^{\text{stable disc}}} + \underbrace{\sum_{k=1}^{\infty} f_{-k} e^{-jk\theta}}_{Z^{\text{unstable disc}}}$$



Step 3: Compute Fourier series

Rational Interpolation



$$M_m(z) = \frac{az^2 + bz + c}{z + d}$$

Continuous derivative
in interpolation points

Quadrature integration

$$f_k = \frac{1}{2\pi} \int_0^{2\pi} Z(\theta) e^{-jk\theta} d\theta$$

- + no extra parameters
- slow

Fast Fourier Transform (FFT)

- + Fast
- #points is extra parameter
- Introduces aliasing

Content

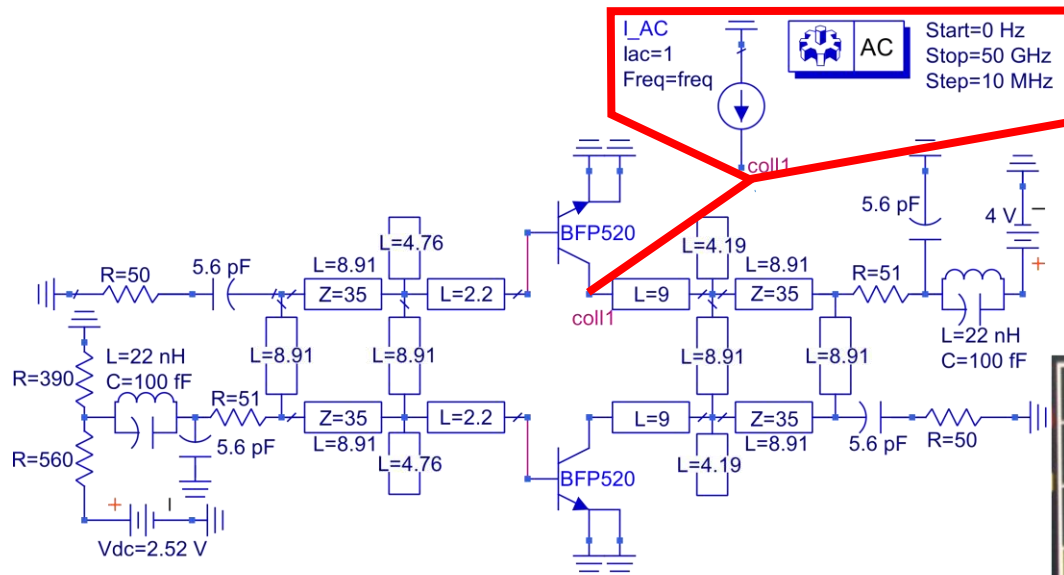
Stability Analysis by projection

Examples

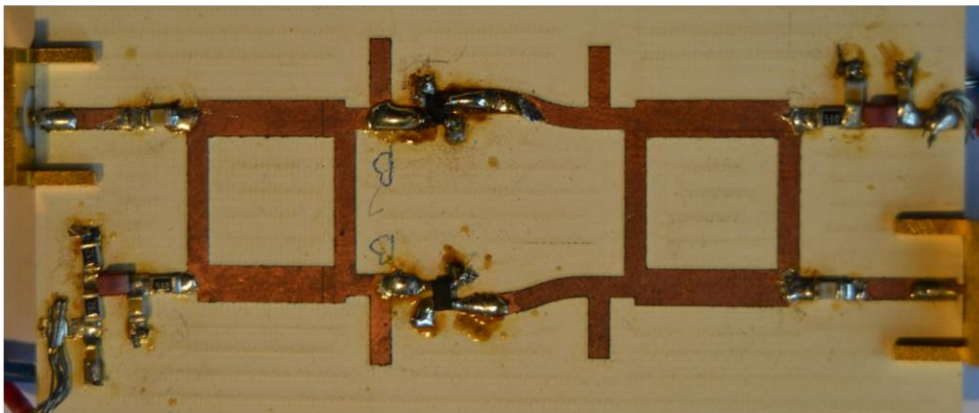
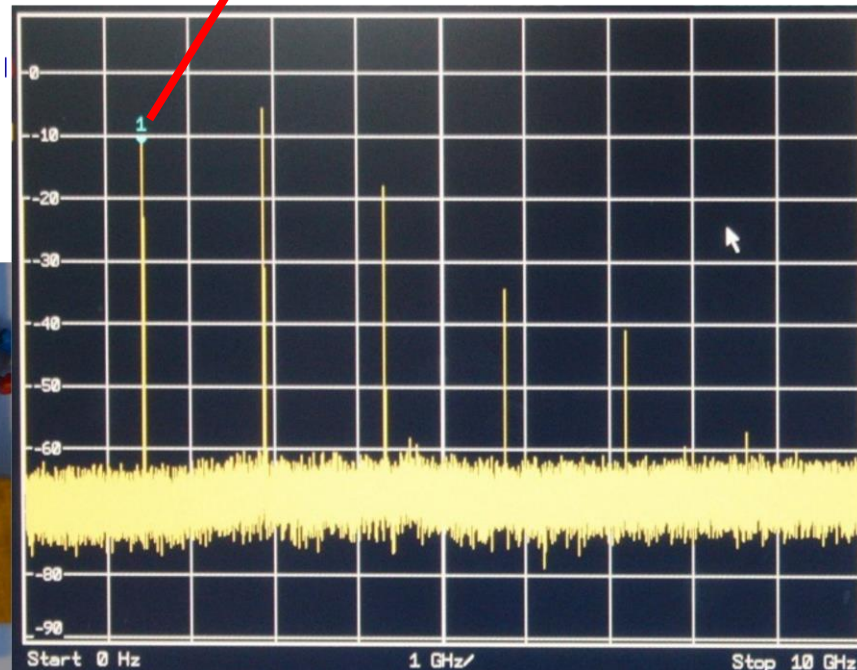
Filter Influence

Estimating unstable poles

Example 1: Balanced Amplifier

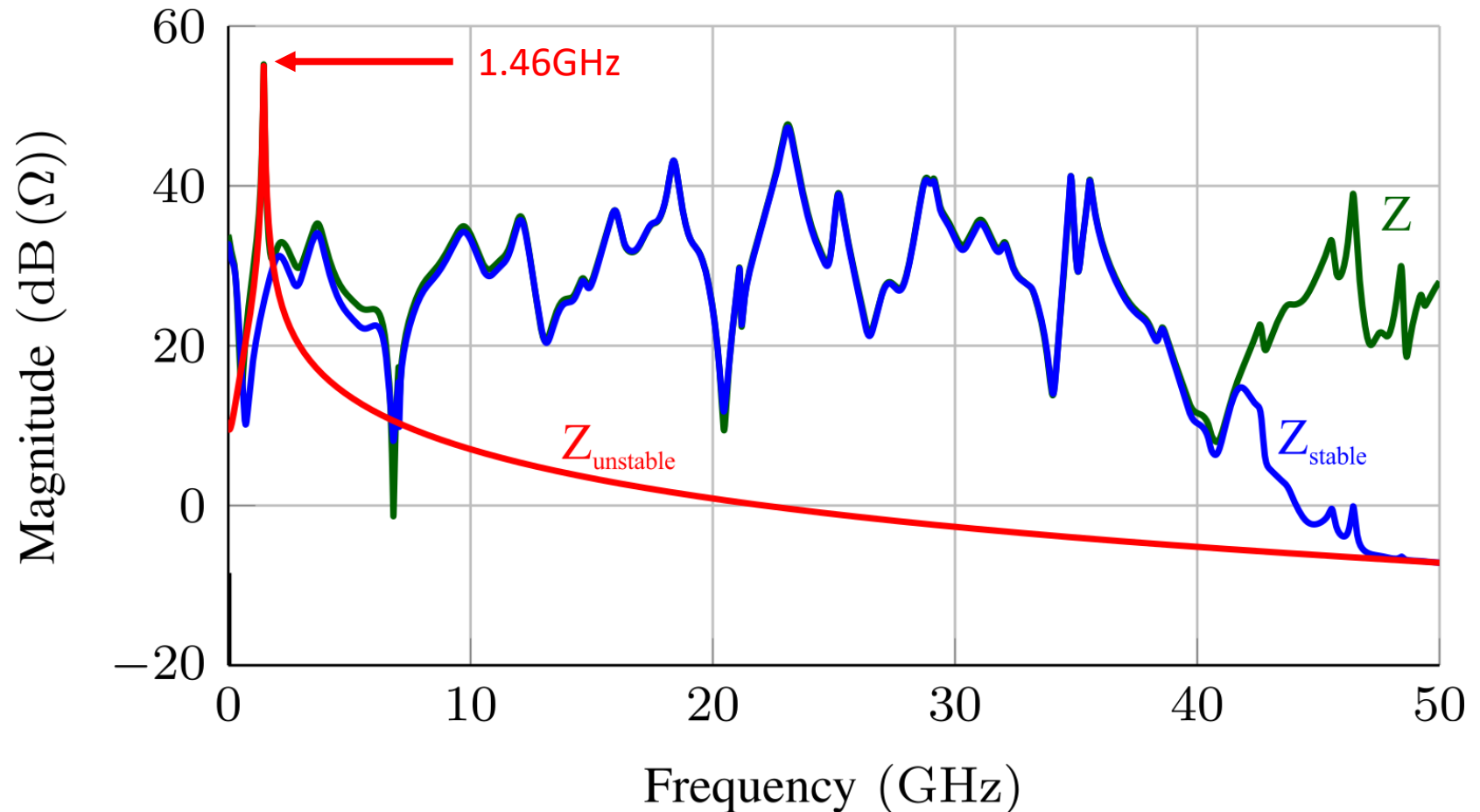


Instability measured
at 1.43GHz



Result: Instability is detected

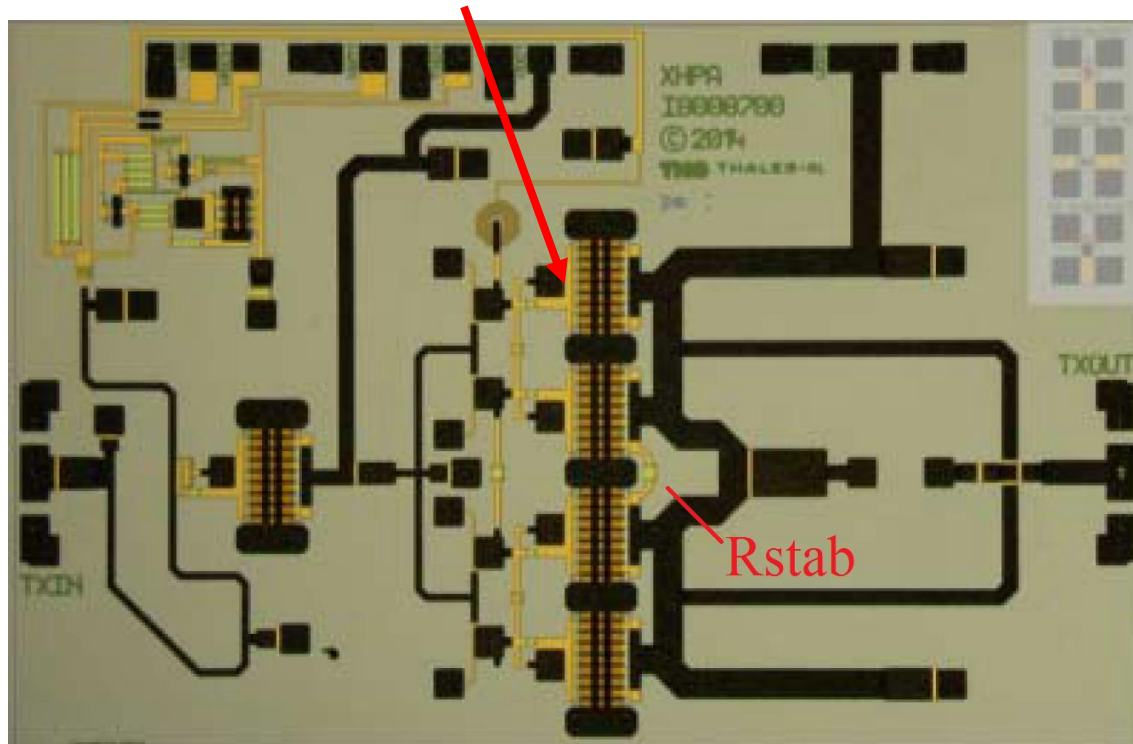
0 – 50GHz in 1MHz steps 50k points. Processing time: 60ms



Example: Power Amplifier

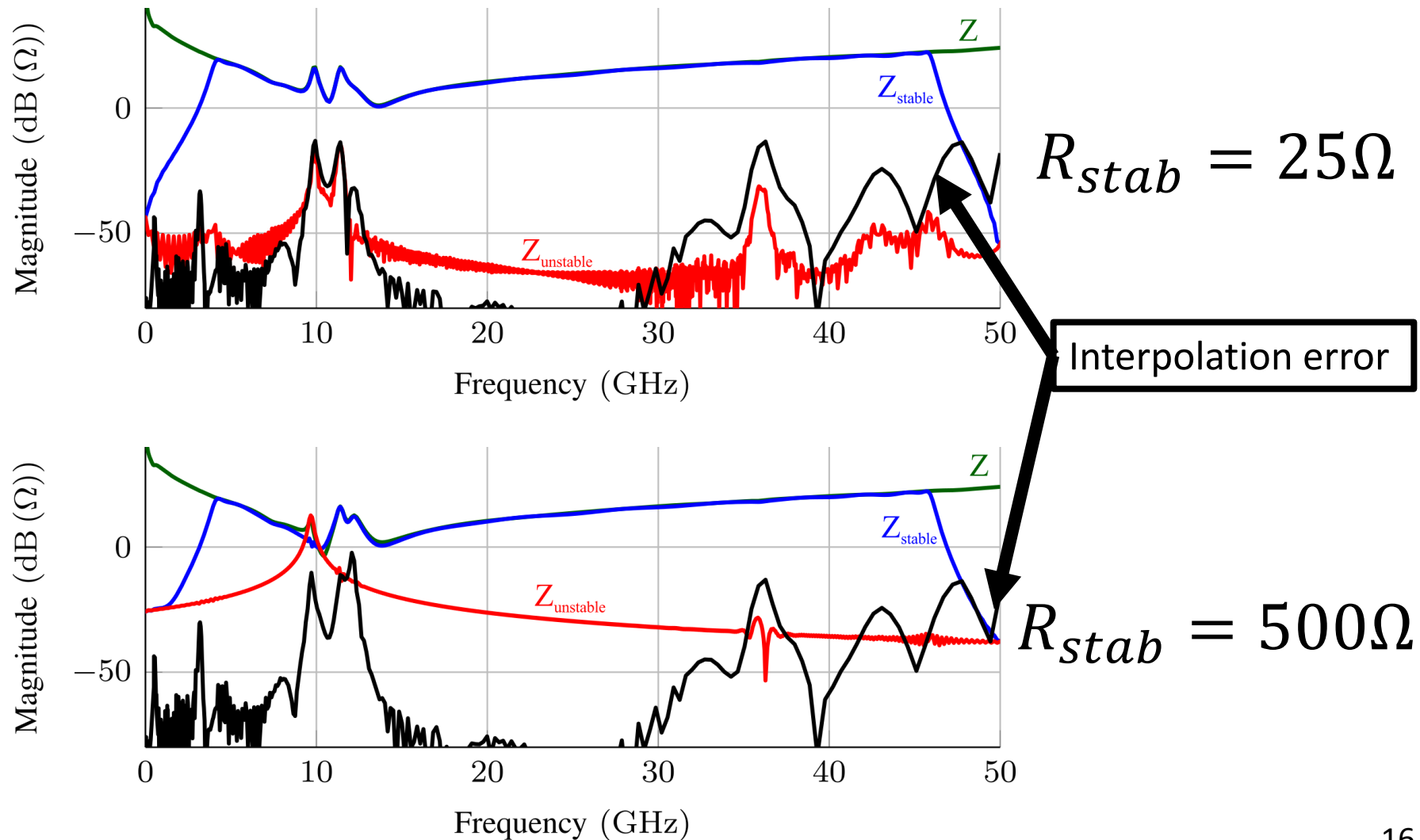
Possible odd-mode instability in second stage

Impedance determined here



Thanks to M. Van Heijningen (TNO) for the simulation data

Results



Content

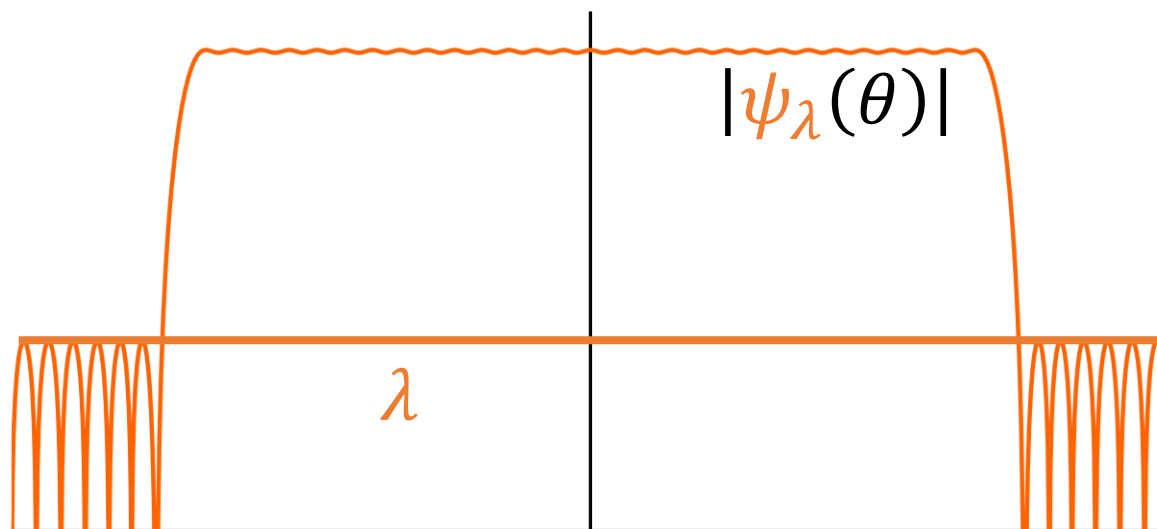
Stability Analysis by projection

Examples

Filter influence

Estimating unstable poles

Influence of the filter



$Z(j\omega)$ unstable pole in γ_i :

$$P_{\bar{\mathcal{H}}_2}\{Z(j\omega)\} = \frac{R_i}{j\omega - \gamma_i} \longrightarrow P_{\bar{\mathcal{H}}_2}\{Z\psi_\lambda\} = \frac{\psi_\lambda(\gamma_i)R_i}{j\omega - \gamma_i}$$

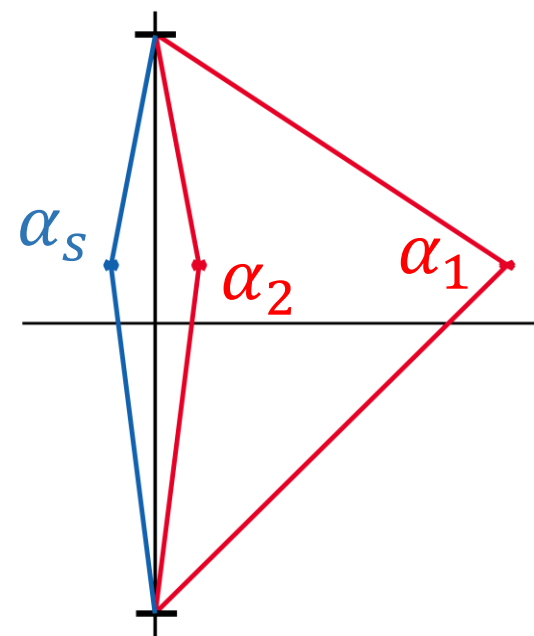
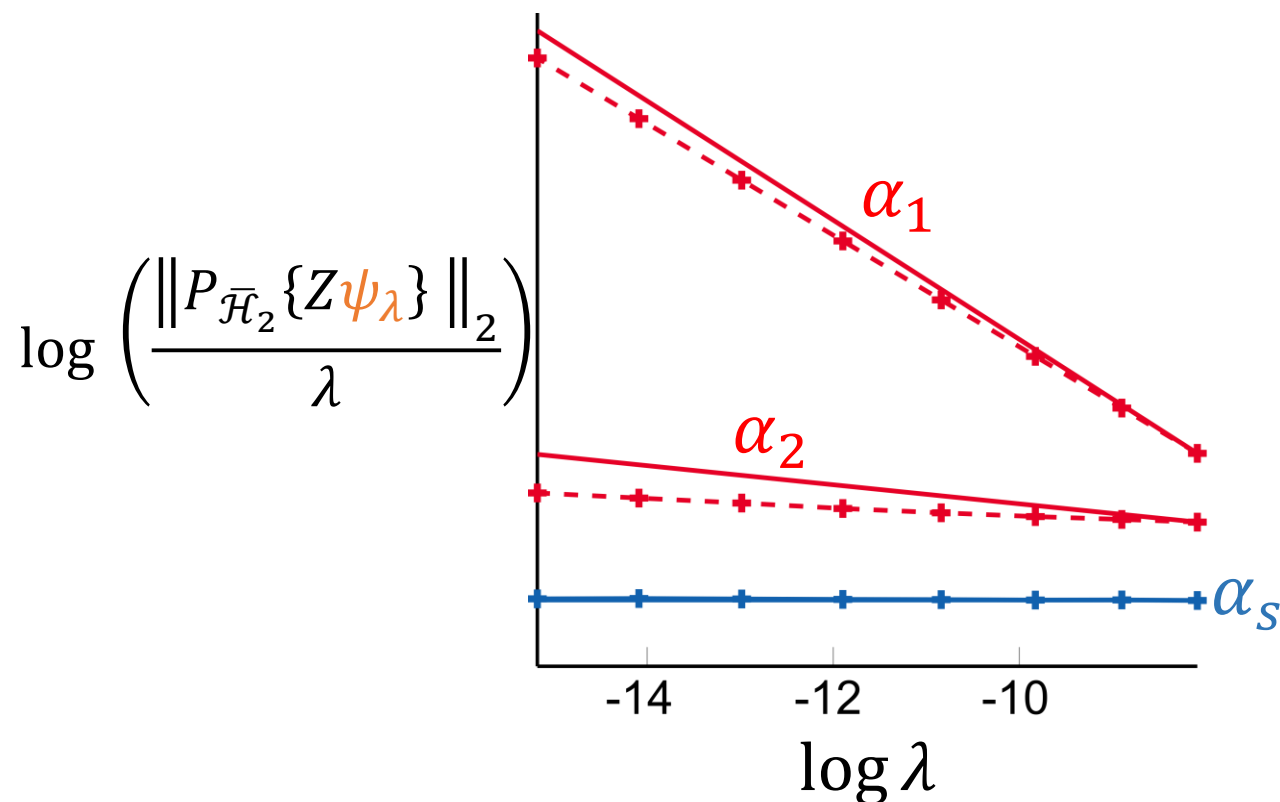
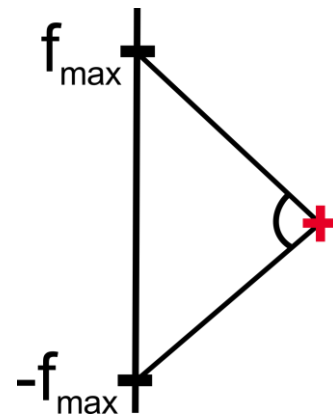
λ too low: might suppress poles

λ too high: influence of edge

Projection as function of λ

Filter magnitude is known:

$$|\psi_\lambda(\gamma_i)| \cong \lambda^{1-\frac{\alpha}{\pi}}$$



Is $Z(j\omega)$ unstable?

Yes or ~~No~~

α very small

Stable OR Unstable pole far away

Projection + λ -analysis:

promising technique (Work in progress)

Content

Stability Analysis by projection

Examples

Filter influence

Estimating unstable poles

Estimating unstable poles

Unstable part = rational

⇒ Classical methods to estimate the poles

Least squares

Levy's method, or more advanced

H_∞ approximation

Adamjan, Arov and Krein (AAK)

 Padé approximation

Padé approx. of unstable part

When system has N poles in the unit circle then

$$\Psi_{N+1} = \begin{pmatrix} f_{-1} & f_{-2} & \cdots & f_{-(N+1)} \\ f_{-2} & f_{-3} & \cdots & f_{-(N+2)} \\ \vdots & \vdots & \ddots & \vdots \\ f_{-(N+1)} & f_{-(N+2)} & \cdots & f_{-(2N+1)} \end{pmatrix}$$

Has rank N

Use singular values of Ψ_M with $M > N$ to determine order

Padé approx. of unstable part

With order N known, compute SVD

$$\Psi_{N+1} = USW'$$

Poles of unstable part are now roots of

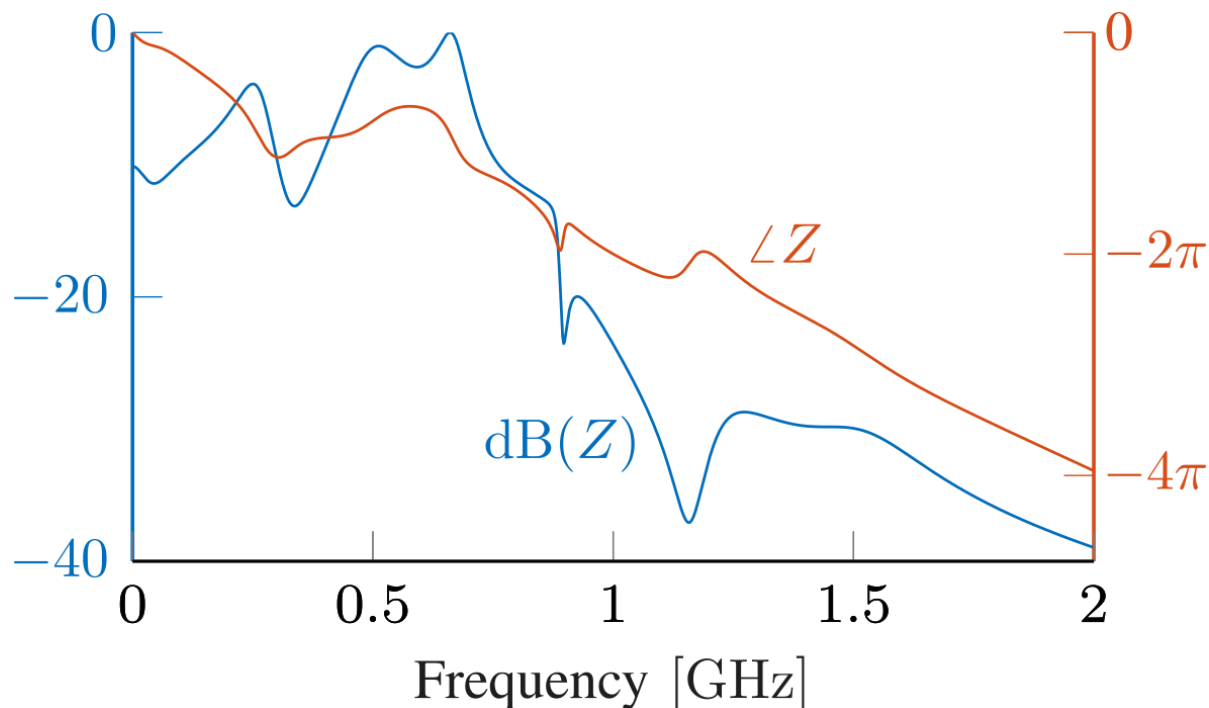
$$W_{N+1,N+1}z^N + W_{N,N+1}z^{N-1} + \dots + W_{1,N+1}$$

Example: random system

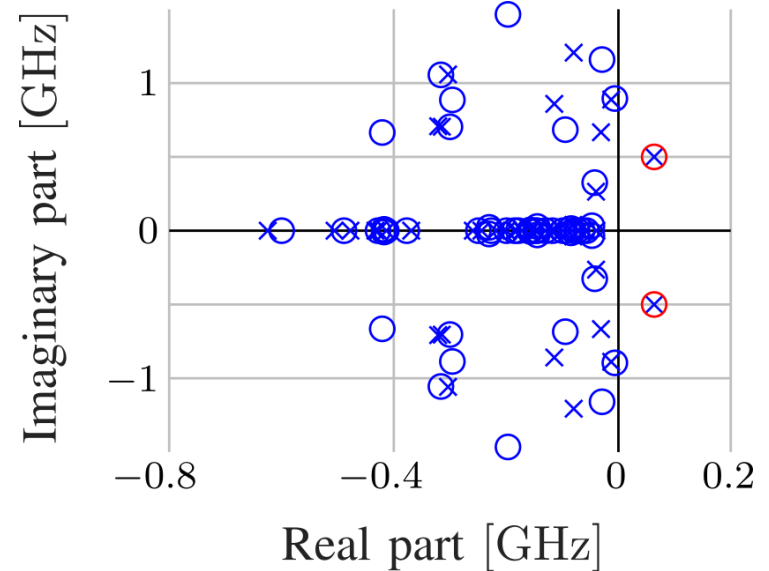
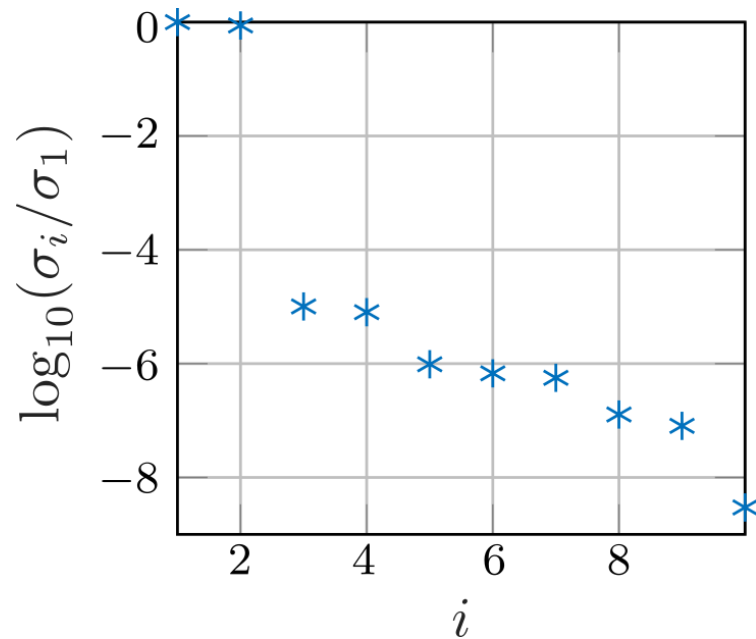
System with 52 poles and 50 zeroes. (RSS Matlab)

Time delay at the input (1ns)

2 unstable poles



Example: random system



$$\text{Error} = \frac{|p_{\text{estimated}} - p_{\text{correct}}|}{|p_{\text{correct}}|} = 4.9 \cdot 10^{-6}$$

Conclusions

Stability analysis with projection

“Model-free” method

Allows non-parametric stab-analysis

Filter influence

Could be used to get automatic yes/maybe answer

Determining unstable poles

Exploit fact that unstable part is rational

Padé approx. only requires small # Fourier coeffs